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A canonical form for the design of unknown input sliding mode observers

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1.1 Introduction

The concept of sliding mode control [11, 27, 32] has been extended to the problem of the state estimation by an observer, both for linear [10, 32] and nonlinear systems [1, 8, 28, 34]. Using the same design theory as variable structure control laws, the observer trajectories are constrained to evolve after a finite time on a suitable sliding manifold, by the use of a discontinuous output injection (the sliding manifold is usually given by the difference between the observer and the system output). The sliding motion provides an estimation (asymptotically or in finite time) of the system state. Sliding mode observers have been shown to be efficient in many applications, such as in robotics [3, 21], electrical engineering [6, 33], chemical reactors [25] or fault detection [14, 18].

The problem of designing an observer for a multivariable linear system partially driven by unknown inputs is of great interest. Such a problem arises in systems subject to disturbances or with inaccessible inputs and in many applications such as fault detection and isolation, parameter identification or cryptography. Since, in the sliding mode, the resulting dynamics are insensitive to a class of perturbations and parametric uncertainties, variable structure techniques for robust state reconstruction have been developed in many papers. Observers are designed using canonical forms which are derived under the assumption that some *matching conditions* are fulfilled by the system.

The main contribution of this chapter is to introduce a constructive algorithm that transforms the system into a new canonical form suitable for the design of finite time sliding mode observers³. An interesting property of the algorithm is that it can be performed to estimate the state of some systems

³ Note that the system has to be fully observable for a finite time convergence property.

that do not necessarily satisfy the classical matching conditions involved in the design of sliding mode observers.

This work is organized as follows. Section 1.2 gives a brief overview of robust state reconstruction based on sliding mode observers and the topic of this work. In Section 1.3 the algorithm is developed, which results in a canonical form made of block triangular observable forms. Then in Section 1.4, the state and the unknown inputs are estimated by means of a finite time observer designed under second order sliding mode considerations. An example, that points out the feasibility and the advantages of the proposed method, is given Section 1.5. Throughout the paper, we use the notation:

$$\Delta(l) = \begin{pmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 \end{pmatrix} \in \mathbb{R}^{l \times l}.$$

1.2 Motivation and problem statement

Motivation:

This chapter is concerned with the design of a sliding mode observer for a linear time-invariant system subject to unknown inputs:

$$\dot{x} = Ax + Bu + Dw \quad (1.1)$$

$$y = [y_1 \cdots y_{p_1}]^T = Cx, \quad y_i = C_i x \quad (1.2)$$

where $x \in \mathbb{R}^n$ is the state vector, $y \in \mathbb{R}^{p_1}$ is the output vector, $u \in \mathbb{R}^q$ represents the known inputs and $w \in \mathbb{R}^m$ stands for the bounded, unknown inputs. A , B , C and D are known constant matrices of appropriate dimension. It is supposed that $m \leq p_1$ and without loss of generality, it can be assumed that $\text{rank}(C) = p_1$ and that $\text{rank}(D) = m$.

The design of asymptotically stable sliding mode observers usually assumes that the system (1.1-1.2) satisfies two structural conditions (see e.g. [11]): the invariant zeros of $\{A, D, C\}$ lie in \mathbb{C}_- and

$$\text{rank}(CD) = \text{rank}(D) = m. \quad (1.3)$$

Condition (1.3) is called the *observer matching condition*, and is the analogue of the well-known *matching condition* [9] for a sliding mode controller to be insensitive to matched perturbations.

Then, there exists a linear change of coordinates that puts the original system into the canonical form given by:

$$\begin{cases} \dot{x}_1 = A_{11}x_1 + A_{12}y + B_1u, \\ \dot{y} = A_{21}x_1 + A_{22}y + B_2u + D_1w(t) \end{cases} \quad (1.4)$$

A simple observer was originally designed in [32]. In [10], the authors proposed a method that utilizes both linear and discontinuous output error injection. In [30], this method was extended with a Linear Matrix Inequality (LMI) formulation in order to exploit additional degrees of freedom in the choice of the linear and discontinuous gains and to set the poles of the linear part of the observer in a specified region of the complex plane. These observers have also been recently used in the applicative field of fault detection and identification [12, 31] where the fault is reconstructed by analyzing the so-called *equivalent output injection* (which is the counterpart of the equivalent control in the design of sliding mode control).

Remark 1. In [34], an observer was developed for linear uncertain systems without any state transformation, but with a constrained Lyapunov problem to be solved. It was shown in [4] that this problem has a solution if and only if the two structural conditions hold.

Remark 2. It is worth noticing that the same conditions are required for the design of linear unknown input observers, that use the conventional Luenberger procedure (see e.g. [22] and [13] for a comparative study of both approaches).

Other works on sliding mode observers deal with finite time state estimation for observable systems. They are based on the equivalent control approach. The system is assumed to be put in a so-called block triangular observable form:

$$\begin{bmatrix} \dot{y} \\ \dot{x}_2 \\ \vdots \\ \dot{x}_{l-1} \\ \dot{x}_l \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} & 0 & 0 & \dots & 0 \\ A_{21} & A_{22} & A_{23} & 0 & \dots & 0 \\ \vdots & & & \ddots & & \vdots \\ \vdots & & & & \ddots & A_{l-1,l} \\ A_{l1} & \dots & & & \dots & A_{l,l} \end{bmatrix} \begin{bmatrix} y \\ x_2 \\ \vdots \\ x_{l-1} \\ x_l \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \\ \vdots \\ B_{l-1} \\ B_l \end{bmatrix} u + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ D_l \end{bmatrix} w \tag{1.5}$$

The estimation of the state is based on a step-by-step procedure using the successive filtered values of equivalent output injections obtained from first order sliding mode observers. One can refer to the works [15, 19, 33] for linear systems or to [1, 7, 8, 35] for nonlinear systems. Applications can be found in fault detection and isolation [16] or in cryptography [2]. Step-by-step first order sliding mode observers were also successfully implemented in electrical and mechanical systems [6, 26].

Such a method is tractable if the unknown inputs appear in the last block of (1.5). For instance, in the simple case of a single output observable system ($p_1 = 1$), the following extended observer matching condition should be satisfied (see [27], Chapter 4):

$$\text{rank} \begin{pmatrix} C \\ CA \\ \vdots \\ CA^{n-2} \end{pmatrix} D = 0. \quad (1.6)$$

Then, under the change of coordinates

$$\xi = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} x,$$

the system (1.1-1.2) is transformed into

$$\begin{aligned} \dot{\xi} &= \Delta(n)\xi + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ CA^n \end{bmatrix} x + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ CA^{n-1}D \end{bmatrix} w + Bu, \\ y &= [1 \ 0 \ \cdots \ 0] \xi. \end{aligned}$$

and a finite time sliding mode observer can be designed as it will be shown in Section 1.4.1.

Problem statement:

Here, the aim is to design a sliding mode observer for system (1.1-1.2), leading to the finite time estimation of both state and unknown inputs, even if the previously mentioned matching conditions (1.3) or (1.6) are not satisfied. To this end, a suitable change of coordinates is provided via a constructive algorithm (as a consequence, this algorithm allows to the assessment of whether full state observability is achieved and the possibility to estimate the unknown inputs). Under this transformation, the system is put in a new type of canonical form. This form is an extension of the classical block triangular observable form, and will be called a *quasi block triangular observable form*. Actually, it is made of a set of block triangular observable forms, but some of them are fed by fictitious outputs. Then, the use of a variable structure observer, and more particularly a step-by-step second order sliding mode observer, ensures the convergence to zero of the observation errors in finite time. The cornerstone of the procedure is the manipulation in a suitable way of all the equivalent output injections, whose analysis leads to the knowledge of the fictitious outputs. Furthermore, this procedure allows the recovery of the unknown inputs in finite time, under the assumption that they are at least bounded piecewise continuous functions.

1.3 Output Information Algorithm

In this Section an algorithm is given that puts the system into a quasi block triangular observable form.

Step 1: Consider the vector of outputs $y^1 \triangleq Cx$.

a. Without loss of generality, reorder the components of y^1 as follows:

$$y^1 = [C_1^T \cdots C_{\eta_1}^T \ C_{\eta_1+1}^T \cdots C_{p_1}^T]^T x$$

so that for $1 \leq j \leq \eta_1$:

$$C_j A^k D = 0, \text{ for all } k \in \mathbb{N},$$

and for $1 \leq j \leq p_1 - \eta_1$, there exists an integer r_j^1 such that:

$$\begin{aligned} C_{\eta_1+j} A^k D &= 0, \text{ for all } k < r_j^1 - 1 \\ C_{\eta_1+j} A^{r_j^1-1} D &\neq 0. \end{aligned}$$

The outputs y_j^1 , $j \leq \eta_1$, are not affected by the unknown inputs.

b. Compute the set of row vectors

$$\Phi^1 = \text{span} \{C_1, \dots, C_1 A^{n-1}, C_2, \dots, C_2 A^{n-1}, \dots, C_{\eta_1}, \dots, C_{\eta_1} A^{n-1}\}$$

and note $\varphi^1 = \text{rank}(\Phi^1)$.

Define the integers φ_j^1 , $1 \leq j \leq \eta_1$, (not uniquely), in such a way that the matrix

$$I_1 = \begin{bmatrix} C_1 \\ \vdots \\ C_1 A^{\varphi_1^1-1} \\ \vdots \\ C_{\eta_1} \\ \vdots \\ C_{\eta_1} A^{\varphi_{\eta_1}^1-1} \end{bmatrix}$$

has rank φ^1 . One has $\varphi_1^1 + \dots + \varphi_{\eta_1}^1 = \varphi^1$. If $\varphi^1 = n$, stop the algorithm. Actually, in such a case, the state is not affected by any disturbance, i.e. $D = 0$.

c. Compute the set of row vectors

$$\Upsilon^1 = \text{span} \{C_{\eta_1+1}, \dots, C_{\eta_1+1} A^{r_1^1-1}, \dots, C_{p_1}, \dots, C_{p_1} A^{r_{p_1-\eta_1}^1-1}\}$$

and write $\varphi^1 + \rho^1 = \text{rank}(\Phi^1 \cup \Upsilon^1)$.

Define the integers $(\rho_1^1, \dots, \rho_{p_1-\eta_1}^1)$ so that (after a possible reordering of the

$$C_i, \eta_1 + 1 \leq i \leq p_1), \text{rank} \begin{bmatrix} I_1 \\ D_1 \end{bmatrix} = \varphi^1 + \rho^1, \text{ where } D_1 = \begin{bmatrix} C_{\eta_1+1} \\ \vdots \\ C_{\eta_1+1} A^{\rho_1^1-1} \\ \vdots \\ C_{p_1} \\ \vdots \\ C_{p_1} A^{\rho_{p_1-\eta_1}^1-1} \end{bmatrix}.$$

One has $\rho_1^1 + \dots + \rho_{p_1-\eta_1}^1 = \rho^1$.

In order to explain here the motivation of such computations, let us set the $(n \times n)$ matrix $T_1 = \begin{bmatrix} I_1 \\ D_1 \\ \Omega_1 \end{bmatrix}$, where $\Omega_1 \in \mathbb{R}^{(n-\varphi^1-\rho^1) \times n}$ is any matrix such

that $\text{rank}(T_1) = n$. Under the change of coordinates $\begin{bmatrix} \sigma^1 \\ \chi^1 \\ \theta^1 \end{bmatrix} = T_1 x$, the system (1.1-1.2) is transformed in the following form:

$$\dot{\sigma}_j^1 = \Delta(\varphi_j^1) \sigma_j^1 + \Xi_j^{\sigma^1} x + B_j^{\sigma^1} u, \quad 1 \leq j \leq \eta_1 \quad (1.7)$$

$$\dot{\chi}_j^1 = \Delta(\rho_j^1) \chi_j^1 + \Xi_j^{\chi^1} x + \Theta_j^{\chi^1} w + B_j^{\chi^1} u, \quad 1 \leq j \leq p_1 - \eta_1 \quad (1.8)$$

$$\dot{\theta}^1 = P_1 \begin{bmatrix} \sigma^1 \\ \chi^1 \\ \theta^1 \end{bmatrix} + Q_1 w + B^{\theta^1} u \quad (1.9)$$

where $\sigma^1 = \begin{bmatrix} \sigma_1^1 \\ \vdots \\ \sigma_{\eta_1}^1 \end{bmatrix}$, $\sigma_j^1 \in \mathbb{R}^{\varphi_j^1}$, and $\chi^1 = \begin{bmatrix} \chi_1^1 \\ \vdots \\ \chi_{p_1-\eta_1}^1 \end{bmatrix}$, $\chi_j^1 \in \mathbb{R}^{\rho_j^1}$, and

$$\Xi_j^{\sigma^1} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_j A^{\varphi_j^1} \end{bmatrix}_{\varphi_j^1 \times n}, \quad \Xi_j^{\chi^1} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_{\eta_1+j} A^{\rho_j^1} \end{bmatrix}_{\rho_j^1 \times n}, \quad \Theta_j^{\chi^1} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_{\eta_1+j} A^{\rho_j^1-1} D \end{bmatrix}_{\rho_j^1 \times m}$$

and $B_j^{\sigma^1}$, $B_j^{\chi^1}$ and B^{θ^1} are matrices of appropriate dimension. The outputs y_j^1 are the first components of σ_j^1 , for $1 \leq j \leq \eta_1$, and of χ_j^1 , for $1 \leq j \leq p_1 - \eta_1$. The subsystem (1.7-1.8) is in a block triangular observable form similar to (1.5). Thus, as argued in Section 1.2, the whole state can be estimated in finite time using existing step-by-step sliding mode observers if $\varphi^1 + \rho^1 = n$. Unfortunately, this condition is not fulfilled by many systems. If $\varphi^1 + \rho^1 < n$,

the main idea of the algorithm is to compute some fictitious outputs⁴ so that the subsystem (1.9) can be put in a quasi block triangular observable form⁵ that allows the estimation of the remaining part of the state θ^1 . For this:

d. Define the matrix

$$\Gamma_1 = \begin{bmatrix} C_{\eta_1+1} A^{r_1^1-1} D \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1-1} D \end{bmatrix}$$

and let $d_1 = \text{rank}(\Gamma_1)$. If $d_1 = p_1 - \eta_1$, stop the algorithm. If $d_1 < p_1 - \eta_1$, there exists a matrix $\Lambda_1 \in \mathbb{R}^{p_2 \times (p_1 - \eta_1)}$, where $p_2 = p_1 - \eta_1 - d_1$, such that $\Lambda_1 \Gamma_1 = 0$. Let us then define the auxiliary variable:

$$y^2 = \Lambda_1 \begin{bmatrix} C_{\eta_1+1} A^{r_1^1} \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1} \end{bmatrix} x \triangleq C^2 x, \quad C^2 = \begin{bmatrix} C_1^2 \\ \vdots \\ C_{p_2}^2 \end{bmatrix}.$$

Note that C^2 is not necessarily full rank.

Step 2: Apply the Output Information Algorithm to the new vector of fictitious outputs $y^2 \in \mathbb{R}^{p_2}$.

a. After possible reordering of the components of y^2 , by analogy with Step 1.a, define the integers η_2 and r_j^2 , $1 \leq j \leq p_2 - \eta_2$.

b. In the same way as Step 1.b, define the set Φ^2 , the integer φ^2 (such that $\text{rank}(\Phi^1 \cup \mathcal{Y}^1 \cup \Phi^2) = \varphi^1 + \rho^1 + \varphi^2$), the integers φ_j^2 , $1 \leq j \leq \eta_2$, and the

related set I_2 such that $\text{rank} \begin{bmatrix} I_1 \\ D_1 \\ I_2 \end{bmatrix} = \varphi^1 + \rho^1 + \varphi^2$. If $\varphi^1 + \rho^1 + \varphi^2 = n$, stop

the algorithm.

c. By analogy with Step 1.c, define the sets \mathcal{Y}^2 and D_2 and the related integers ρ^2 and $(\rho_1^2, \dots, \rho_{p_2-\eta_2}^2)$.

If $\varphi^1 + \rho^1 + \varphi^2 + \rho^2 = n$, or if $\{\varphi^1 + \rho^1 + \varphi^2 + \rho^2 < n \text{ and } D_2 = \emptyset\}$, stop the algorithm. The second condition means that there is no fictitious output (affected by the unknown inputs) linearly independent from the previous measurements, that could lead to a (even partially) block triangular observable form for the subsystem (1.9).

d. Define the matrix

$$\Gamma_2 = \begin{bmatrix} \Gamma_1 \\ C_{\eta_2+1}^2 A^{r_1^2-1} D \\ \vdots \\ C_{p_2}^2 A^{r_{p_2-\eta_2}^2-1} D \end{bmatrix}$$

⁴ these fictitious outputs will be made available by analyzing the equivalent output injections of a suitable sliding mode observer, see Section 1.4.

⁵ the term *quasi* is used because some blocks will be fed by fictitious outputs and not by measured outputs.

and let $d_2 = \text{rank } \Gamma_2$. If $d_2 < (p_1 - \eta_1) + (p_2 - \eta_2)$, one can find a matrix $\Lambda_2 \in \mathbb{R}^{p_3 \times ((p_1 - \eta_1) + (p_2 - \eta_2))}$, where $p_3 = (p_1 - \eta_1) + (p_2 - \eta_2) - d_2$, such that $\Lambda_2 \Gamma_2 = 0$. Then the Output Information Algorithm is applied to the new fictitious outputs

$$y^3 = \Lambda_2 \begin{bmatrix} C_{\eta_1+1} A^{r_1^1} \\ \vdots \\ C_{p_1} A^{r_{p_1-1}^1} \\ C_{\eta_2+1}^2 A^{r_1^2} \\ \vdots \\ C_{p_2}^2 A^{r_{p_2-\eta_2}^2} \end{bmatrix} x \triangleq C^3 x.$$

Repeating this procedure, one has:

Step k: Consider the fictitious output $y^k \in \mathbb{R}^{p_k}$, defined in Step $k-1$.

- a. Define the integers η_k and r_j^k , $1 \leq j \leq p_k - \eta_k$.
- b. Compute the set of row vectors

$$\Phi^k = \text{span} \{ C_1^k, \dots, C_1^k A^{n-1}, C_2^k, \dots, C_2^k A^{n-1}, \dots, C_{\eta_k}^k, \dots, C_{\eta_k}^k A^{n-1} \}$$

and write $\sum_{i=1}^{k-1} (\varphi^i + \rho^i) + \varphi^k = \text{rank} \left(\left(\bigcup_{i=1}^{k-1} \Phi^i \cup \Upsilon^i \right) \cup \Phi^k \right)$.

Define the integers φ_j^k , $1 \leq j \leq \eta_k$ such that

$$\text{rank} \begin{bmatrix} I_1 \\ D_1 \\ \vdots \\ I_k \end{bmatrix} = \sum_{i=1}^{k-1} (\varphi^i + \rho^i) + \varphi^k$$

where $I_k = \left[(C_1^k)^T, \dots, (C_1^k A^{\varphi_1^k-1})^T, \dots, (C_{\eta_k}^k)^T, \dots, (C_{\eta_k}^k A^{\varphi_{\eta_k}^k-1})^T \right]^T$. If

$\sum_{i=1}^{k-1} (\varphi^i + \rho^i) + \varphi^k = n$, stop the algorithm.

- c. Compute the set of row vectors

$$\Upsilon^k = \text{span} \{ C_{\eta_k+1}^k, \dots, C_{\eta_k+1}^k A^{r_1^k-1}, \dots, C_{p_k}^k, \dots, C_{p_k}^k A^{r_{p_k-\eta_k}^k-1} \}$$

and write $\text{rank} \left(\bigcup_{i=1}^k \Phi^i \cup \Upsilon^i \right) = \sum_{i=1}^k (\varphi^i + \rho^i)$.

Define the integers $(\rho_1^k, \dots, \rho_{p_k-\eta_k}^k)$ in such a way that

$$\text{rank} \begin{bmatrix} I_1 \\ D_1 \\ \vdots \\ I_k \\ D_k \end{bmatrix} = \sum_{i=1}^k (\varphi^i + \rho^i)$$

where

$$D_k = \left[(C_{\eta_{k+1}}^k)^T, \dots, (C_{\eta_{k+1}}^k A^{\rho_1^k - 1})^T, \dots, (C_{p_k}^k)^T, \dots, (C_{p_k}^k A^{\rho_{p_k}^k - \eta_k - 1})^T \right]^T.$$

If $\sum_{i=1}^k (\varphi^i + \rho^i) = n$, or if $\left\{ \sum_{i=1}^k (\varphi^i + \rho^i) < n \text{ and } D_k = \emptyset \right\}$, stop the algorithm.

d. Define

$$\Gamma_k = \begin{bmatrix} \Gamma_{k-1} \\ C_{\eta_{k+1}}^k A^{r_1^k - 1} D \\ \vdots \\ C_{p_k}^k A^{r_{p_k}^k - \eta_k - 1} D \end{bmatrix}$$

and let $d_k = \text{rank}(\Gamma_k)$. If $d_k < \sum_{s=1}^k (p_s - \eta_s)$, let us set $p_{k+1} = \sum_{s=1}^k (p_s - \eta_s) -$

d_k . There exists a matrix $\Lambda_k \in \mathbb{R}^{p_{k+1} \times \sum_{s=1}^k (p_s - \eta_s)}$ such that $\Lambda_k \Gamma_k = 0$. Then a new fictitious output is defined:

$$y^{k+1} = \Lambda_k \begin{bmatrix} C_{\eta_1+1} A^{r_1^1} \\ \vdots \\ C_{p_1} A^{r_{p_1}^1 - \eta_1} \\ \vdots \\ C_{\eta_k+1} A^{r_1^k} \\ \vdots \\ C_{p_k} A^{r_{p_k}^k - \eta_k} \end{bmatrix} x \triangleq C^{k+1} x.$$

Stop the algorithm if:

1. there exists a $\mu \in \mathbb{N}$, such that

$$\sum_{i=1}^{\mu} (\varphi^i + \rho^i) < n \text{ and } \left\{ d_{\mu} = \sum_{s=1}^{\mu} (p_s - \eta_s) \text{ or } D_{\mu} = \emptyset \right\},$$

2. there exists a $k^* \in \mathbb{N}$ such that $\sum_{i=1}^{k^*} (\varphi^i + \rho^i) = n$.

One has $\mu, k^* \leq n - p_1$. So, the algorithm stops after a number of steps less than $n - p_1$. In case 1, it is not possible to find new variables leading to a quasi block triangular observable form. In case 2, one can set the following

nonsingular $(n \times n)$ matrix $T_{k^*} = \begin{bmatrix} I_1 \\ D_1 \\ \vdots \\ I_{k^*} \\ D_{k^*} \end{bmatrix}$. Under the coordinate transfor-

mation $\begin{bmatrix} \sigma^1 \\ \chi^1 \\ \vdots \\ \sigma^{k^*} \\ \chi^{k^*} \end{bmatrix} = T_{k^*} x$ where, for $1 \leq i \leq k^*$, $\sigma^i = \begin{bmatrix} \sigma_1^i \\ \vdots \\ \sigma_{\eta_i}^i \end{bmatrix}$, $\sigma_j^i \in \mathbb{R}^{\varphi_j^i}$, and

$\chi^i = \begin{bmatrix} \chi_1^i \\ \vdots \\ \chi_{p_i - \eta_i}^i \end{bmatrix}$, $\chi_j^i \in \mathbb{R}^{\rho_j^i}$, the system (1.1-1.2) becomes:

$$\dot{\sigma}_j^i = \Delta(\varphi_j^i) \sigma_j^i + \Xi_j^{\sigma^i} x + B_j^{\sigma^i} u, \quad 1 \leq j \leq \eta_i \quad (1.10)$$

$$\dot{\chi}_j^i = \Delta(\rho_j^i) \chi_j^i + \Xi_j^{\chi^i} x + \Theta_j^{\chi^i} w + B_j^{\chi^i} u, \quad 1 \leq j \leq p_i - \eta_i \quad (1.11)$$

$$\Xi_j^{\sigma^i} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_j A^{\varphi_j^i} \end{bmatrix}_{\varphi_j^i \times n}, \quad \Xi_j^{\chi^i} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_{\eta_i+j} A^{\rho_j^i} \end{bmatrix}_{\rho_j^i \times n}, \quad \Theta_j^{\chi^i} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ C_{\eta_i+j} A^{\rho_j^i-1} D \end{bmatrix}_{\rho_j^i \times m}.$$

$B_{i,j}^{\sigma^i}$ and $B_{i,j}^{\chi^i}$ are a $(\varphi_j^i \times q)$ and a $(\rho_j^i \times q)$ -matrix, respectively. Thus, the system is put in quasi block triangular observable form.

1.4 Observer design

1.4.1 Sliding mode observer for a single triangular observable form

In this part, a step-by-step sliding mode observer is designed for a system described by the following triangular form:

$$\dot{\xi} = \Delta(l)\xi + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ M \end{bmatrix} z + Bu, \quad y = [1 \ 0 \ \cdots \ 0] \xi \quad (1.12)$$

where $\xi = [\xi_1 \ \cdots \ \xi_l]^T \in \mathbb{R}^l$, $(l > 1)$ is the state vector, $y \in \mathbb{R}$ is the output vector, u is the input vector and $z \in \mathbb{R}^m$ stands for some state variables and unknown inputs with $M \in \mathbb{R}^{1 \times m}$. Let us assume that the system is

Bounded Input Bounded State (BIBS) and that z and its first time derivative are bounded, i.e.:

$$\begin{aligned} |\xi_i| &< d_i, \quad i = 1, \dots, l \\ \|z\| &< K \\ \|\dot{z}\| &< K', \end{aligned}$$

where d_i , K and K' are some known positive scalars.

Most of sliding mode observer designs for (1.12) are based on a step-by-step procedure using successive filtering values of the so-called equivalent output injections obtained from recursive first order sliding mode observers. However, the approximation of the equivalent information injections by low pass filters at each step may introduce some delays that could lead to inaccurate estimates or to instability for high order systems⁶. To overcome this problem, here it is proposed to replace the discontinuous first order sliding mode output injection by a continuous second order sliding mode one. The observer is built as follows:

$$\begin{cases} \frac{d\hat{\xi}_1}{dt} = \nu (y - \hat{\xi}_1) \\ \frac{d\hat{\xi}_2}{dt} = E_1 \nu (\tilde{\xi}_2 - \hat{\xi}_2) \\ \vdots \\ \frac{d\hat{\xi}_{l-1}}{dt} = E_{l-2} \nu (\tilde{\xi}_{l-1} - \hat{\xi}_{l-1}) \\ \frac{d\hat{\xi}_l}{dt} = E_{l-1} \nu (\tilde{\xi}_l - \hat{\xi}_l) \end{cases} + Bu, \quad (1.13)$$

with

$$\begin{aligned} \tilde{\xi}_1 &= y \\ \tilde{\xi}_j &= \nu (\tilde{\xi}_{j-1} - \hat{\xi}_{j-1}), \quad 2 \leq j \leq l \end{aligned}$$

where the continuous output injection ν is given by the so-called super twisting algorithm [17, 23]:

$$\begin{cases} \nu(s) = \varphi(s) + \lambda_s |s|^{\frac{1}{2}} \text{sign}(s) \\ \dot{\varphi}(s) = \alpha_s \text{sign}(s) \\ \lambda_s, \alpha_s > 0 \end{cases}. \quad (1.14)$$

Conditions on choice of λ_s and α_s will be explained later. The function E_i is equal to 1 if all the previous error components have converged to zero, else $E_i = 0$. This is an anti-peaking structure [29]. Indeed, with this particular function, the manifolds are reached one by one. At each step, a subdynamics of dimension one is obtained and consequently no peaking phenomena appear. Denoting $e = \xi - \hat{\xi}$, the error dynamics is given by:

⁶ In [20], the implementation of such filters was studied and a relation between the estimation accuracy, the filter time constants and the sampling time was given.

$$\left\{ \begin{array}{l} \dot{e}_1 = \xi_2 - \nu(y - \hat{\xi}_1) \\ \dot{e}_2 = \xi_3 - E_1 \nu(\tilde{\xi}_2 - \hat{\xi}_2) \\ \vdots \\ \dot{e}_{l-1} = \xi_l - E_{l-2} \nu(\tilde{\xi}_{l-1} - \hat{\xi}_{l-1}) \\ \dot{e}_l = Mz - E_{l-1} \nu(\tilde{\xi}_l - \hat{\xi}_l) \end{array} \right. \quad (1.15)$$

First step: Before sliding mode on $e_1 = 0$, one has:

$$\left\{ \begin{array}{l} \dot{e}_1 = \xi_2 - \nu(e_1) \\ \dot{e}_2 = \xi_3 \\ \vdots \\ \dot{e}_{l-1} = \xi_l \\ \dot{e}_l = Mz \end{array} \right.$$

The super twisting algorithm has been developed for systems with relative degree one with respect to the input as in $\dot{e}_1 = \xi_2 - \nu(e_1)$. The second time derivative of e_1 is given by:

$$\ddot{e}_1 = \dot{\xi}_2 - \dot{\nu}(e_1) = \xi_3 - \frac{1}{2} \lambda_{e_1} \dot{e}_1 |e_1|^{\frac{1}{2}} - \alpha_{e_1} \text{sign}(e_1).$$

Note that the discontinuity is embedded in the higher derivatives of the error variables, so that they are less affected by the chattering phenomenon. It was shown in [24] that sufficient conditions for the finite time convergence on the second order sliding set $\{e_1 = \dot{e}_1 = 0\}$ are:

$$\begin{aligned} \alpha_{e_1} &> d_3, \\ \lambda_{e_1}^2 &> 4d_3 \frac{\alpha_{e_1} + d_3}{\alpha_{e_1} - d_3}. \end{aligned}$$

Then, the equivalent dynamics provides a continuous estimation (without any introduction of low pass filters) in finite time of ξ_2 since $\nu(e_1) = \xi_2$ on $\dot{e}_1 = 0$. This also implies that $\tilde{\xi}_2 - \hat{\xi}_2 = e_2$.

Second step: after a finite time T_1 , (1.15) becomes:

$$\left\{ \begin{array}{l} \dot{e}_1 = 0 \\ \dot{e}_2 = \xi_3 - \nu(e_2) \\ \vdots \\ \dot{e}_{l-1} = \xi_l \\ \dot{e}_l = Mz \end{array} \right.$$

and one gets the finite time estimation of ξ_3 .

Following the same scheme till the $(l-1)$ -th step, the observer (1.13) provides an estimation of the whole state ξ in finite time.

l-th **step**: (1.15) is given by:

$$\begin{cases} \dot{e}_1 = \dots = \dot{e}_{l-1} = 0 \\ \dot{e}_l = Mz - \nu(e_l) \end{cases}$$

Thus, with the following choice of gains,

$$\begin{aligned} \alpha_{e_l} &> \|M\| K', \\ \lambda_{e_l}^2 &> 4 \|M\| K' \frac{\alpha_{e_l} + \|M\| K'}{\alpha_{e_l} - \|M\| K'}, \end{aligned}$$

one obtains directly, on the sliding manifold $e_l = \dot{e}_l = 0$, a continuous approximation of the unknown variables:

$$\nu(e_l) = Mz$$

The step-by-step observer achieves finite time convergence of the state components. This property is often desirable in the framework of observation, fault detection or identification problems, and in general to solve in finite time the problem of left inversion and state observation.

1.4.2 A finite time observer for the quasi block triangular observable form

It can be seen that, after transformation, the system (1.10)-(1.11) is made of a set of triangular forms similar to (1.12). Thus, for $1 \leq i \leq k^*$, the observer is described by:

$$\frac{d}{dt} \hat{\sigma}_j^i = \Xi_j^{\sigma^i} \hat{x} + B_j^{\sigma^i} u + \bar{\nu}(y_j^i, \hat{\sigma}_j^i, \tilde{\sigma}_j^i), \quad 1 \leq j \leq \eta_i \quad (1.16)$$

$$\frac{d}{dt} \hat{\chi}_j^i = \Xi_j^{\chi^i} \hat{x} + B_j^{\chi^i} u + \bar{\nu}(y_{\eta_i+j}^i, \hat{\chi}_j^i, \tilde{\chi}_j^i), \quad 1 \leq j \leq p_i - \eta_i \quad (1.17)$$

where $\hat{x} = T_{k^*}^{-1} \begin{bmatrix} \hat{\sigma}^1 \\ \hat{\chi}^1 \\ \vdots \\ \hat{\sigma}^{k^*} \\ \hat{\chi}^{k^*} \end{bmatrix}$, and where

$$\bar{\nu}(y_j^i, \hat{\sigma}_j^i, \tilde{\sigma}_j^i) = \begin{bmatrix} \nu(y_j^i - \hat{\sigma}_{j,1}^i) \\ E_2 \nu(\tilde{\sigma}_{j,2}^i - \hat{\sigma}_{j,2}^i) \\ \vdots \\ E_{\varphi_j^i} \nu(\tilde{\sigma}_{j,\varphi_j^i}^i - \hat{\sigma}_{j,\varphi_j^i}^i) \end{bmatrix}$$

and

$$\bar{\nu}(y_{\eta_i+j}^i, \hat{\chi}_j^i, \tilde{\chi}_j^i) = \begin{bmatrix} \nu(y_{\eta_i+j}^i - \hat{\chi}_{j,1}^i) \\ E_2 \nu(\tilde{\chi}_{j,2}^i - \hat{\chi}_{j,2}^i) \\ \vdots \\ E_{\rho_j^i} \nu(\tilde{\chi}_{j,\rho_j^i}^i - \hat{\chi}_{j,\rho_j^i}^i) \end{bmatrix}.$$

$$\begin{aligned} \tilde{\sigma}_{j,1}^i &= y_j^i \\ \tilde{\sigma}_{j,k}^i &= \nu(\tilde{\sigma}_{j,k-1}^i - \hat{\sigma}_{j,k-1}^i) \text{ for } k = 2, \dots, \varphi_j^i, \\ \tilde{\chi}_{j,1}^i &= y_{\eta_i+j}^i \\ \tilde{\chi}_{j,k}^i &= \nu(\tilde{\chi}_{j,k-1}^i - \hat{\chi}_{j,k-1}^i) \text{ for } k = 2, \dots, \rho_j^i. \end{aligned}$$

$\hat{\sigma}_{j,k}^1$ and $\hat{\chi}_{j,k}^1$ are the k -th components of $\hat{\sigma}_j^1$ and $\hat{\chi}_j^1$, respectively and ν is the continuous output injection defined in (1.14). The following hypothesis are required:

- H1. *the system is BIBS: Bounded Input Bounded State,*
- H2. *the term w is bounded and is Lebesgue integrable.*

The following proposition summarizes the results given in this work:

Proposition 1. *Assume that there exists a $k^* \in \mathbb{N}$ such that $\sum_{i=1}^{k^*} (\varphi^i + \rho^i) = n$ and that hypothesis H1 and H2 are satisfied. Then:*

- (i) *rank(Γ_{k^*}) = m ,*
- (ii) *the state is observable and the unknown inputs can be estimated, both of them in finite time by using the step-by-step sliding mode observer (1.16)-(1.17).*

Proof:

- (i) From the definitions of I_i and D_i , and since $\sum_{i=1}^{k^*} (\varphi^i + \rho^i) = n$:

$$\begin{aligned} \text{rank}(\Gamma_{k^*}) &= \text{rank} \begin{bmatrix} C_{\eta_1+1} A_1^{r_1^1-1} \\ \vdots \\ C_{p_1} A_1^{r_1^1-\eta_1-1} \\ \vdots \\ C_{\eta_{k^*}+1} A_1^{r_1^{k^*}-1} \\ \vdots \\ C_{p_{k^*}} A_1^{r_1^{k^*}-\eta_{k^*}-1} \end{bmatrix} D \\ &= \text{rank} \begin{bmatrix} I_1 \\ D_1 \\ \vdots \\ I_{k^*} \\ D_{k^*} \end{bmatrix} D = m. \end{aligned}$$

(ii) The convergence of the observer is proved hereafter.

Convergence of the state variables

We first design an observer for the subsystems given by $i = 1$, i.e. with the available measurement vector $y_j^1 = C_j x$, $1 \leq j \leq p_1$. According to (1.16-1.17), the error dynamics $\bar{\sigma}^1 = \sigma^1 - \hat{\sigma}^1$ and $\bar{\chi}^1 = \chi^1 - \hat{\chi}^1$ are described by the following differential equations:

$$\frac{d}{dt} \bar{\sigma}_j^1 = \Delta(\varphi_j^1) \sigma_j^1 + \Xi_j^{\sigma^1} (x - \hat{x}) - \bar{\nu} (y_j^1, \hat{\sigma}_j^1, \tilde{\sigma}_j^1), \quad (1.18)$$

$$\frac{d}{dt} \bar{\chi}_j^1 = \Delta(\rho_j^1) \chi_j^1 + \Xi_j^{\chi^1} (x - \hat{x}) + \Theta_j^{\chi^1} w - \bar{\nu} (y_{\eta_1+j}^1, \hat{\chi}_j^1, \tilde{\chi}_j^1). \quad (1.19)$$

Every subsystem in (1.18) and (1.19) is in a form similar to (1.15). Thus the variables $\bar{\sigma}^1$ and $\bar{\chi}^1$ tend to zero in a finite time t_1 . Particularly, a sliding mode occurs on the manifolds $\bar{\chi}_{j,\rho_j^1}^1 = 0$, $1 \leq j \leq p_1 - \eta_1$, and the analysis of the resulting equivalent dynamics on the sliding surfaces (obtained by writing that $\frac{d}{dt} \bar{\chi}_{j,\rho_j^1}^1 = 0$) provides the following $(p_1 - \eta_1)$ algebraic equations:

$$\nu \left(\bar{\chi}_{j,\rho_j^1}^1 \right) = C_{\eta_1+j} A^{\rho_j^1} (x - \hat{x}) + C_{\eta_1+j} A^{\rho_j^1-1} D w. \quad (1.20)$$

It can be noted that in the subsequent subsystems ($2 \leq i \leq k^*$), the information injections are not directly available since they are linear combination of some unknown states. Nevertheless, the equivalent control methodology can be used to obtain additional information about y^2 from equations (1.20). Let us introduce the auxiliary variable

$$\tilde{y}^1 = A_1 \begin{bmatrix} C_{\eta_1+1} A^{r_1^1-1} \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1-1} \end{bmatrix} x. \quad (1.21)$$

Note that from the construction of Υ^1 , I_1 and D_1 in Step 1.c. of the algorithm, one can write that

$$\begin{bmatrix} C_{\eta_1+1} A^{r_1^1-1} \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1-1} \end{bmatrix} = G_1 \begin{bmatrix} I_1 \\ D_1 \end{bmatrix}$$

where G_1 is a matrix of appropriate dimension. Thus \tilde{y}^1 can be rewritten as

$$\tilde{y}^1 = A_1 G_1 \begin{bmatrix} I_1 \\ D_1 \end{bmatrix} x = A_1 G_1 \begin{bmatrix} \sigma^1 \\ \chi^1 \end{bmatrix}$$

and represents available information. Then, the following second order sliding mode observer can be designed

$$\frac{d\zeta^1}{dt} = \nu(\tilde{y}^1 - \zeta^1).$$

The dynamics of the error $\varepsilon = \tilde{y}^1 - \zeta^1$ is given by

$$\begin{aligned} \dot{\varepsilon} &= A_1 \begin{bmatrix} C_{\eta_1+1} A^{r_1^1-1} \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1-1} \end{bmatrix} (Ax + Dw) - \nu(\tilde{y}^1 - \zeta^1) \\ &= C^2 x - \nu(\varepsilon). \end{aligned}$$

Thus, with a suitable choice of the gains of the output injection $\nu(\varepsilon)$, a sliding mode occurs on, $\varepsilon = \dot{\varepsilon} = 0$. Then the fictitious output is known after a finite time since:

$$\nu(\varepsilon) = C^2 x = y^2.$$

Remark 3. If $\rho_j^1 = r_j^1$ for all $1 \leq j \leq p_1 - \eta_1$, y^2 can be obtained without any additional observer. Indeed, equations (1.20) become:

$$\begin{aligned} \nu^1 &= \begin{bmatrix} \nu(\bar{\chi}_{1,\rho_1^1}^1) \\ \vdots \\ \nu(\bar{\chi}_{p_1-\eta_1,\rho_{p_1-\eta_1}^1}^1) \end{bmatrix} \\ &= \begin{bmatrix} C_{\eta_1+1} A^{r_1^1} \\ \vdots \\ C_{p_1} A^{r_{p_1-\eta_1}^1} \end{bmatrix} (x - \hat{x}) + \Gamma_1 w. \end{aligned}$$

Since $\Lambda_1 \Gamma_1 = 0$ and from the definition of y^2 , one has:

$$\Lambda_1 \nu^1 = y^2 - \hat{y}^2.$$

The fictitious output y^2 is henceforth available and repeating the same procedure, for $i = 2$, in the observer (1.16-1.17), one obtains an estimation of σ^2 and χ^2 after a finite time t_2 .

Following the same scheme step by step, one gets σ^i and χ^i , $1 \leq i \leq k^*$ in a finite time t_{k^*} . Thus the whole state of the system has been observed.

Estimation of the unknown inputs

After t_{k^*} , additional information can be collected from the equivalent dynamics on the sliding surfaces, namely all the equivalent information injections, for $1 \leq i \leq k^*$:

$$\nu(\bar{\chi}_{j,\rho_j^i}^i) = C_{\eta_i+j} A^{\rho_j^i-1} Dw, \quad 1 \leq j \leq p_i - \eta_i. \quad (1.22)$$

This can be written in compact form $V = \Theta^D w$, where

$$\Theta^D = \begin{bmatrix} \Theta_1^D \\ \vdots \\ \Theta_{k^*}^D \end{bmatrix}, \Theta_i^D = \begin{bmatrix} C_{\eta_i+1} A^{\rho_i^i-1} D \\ \vdots \\ C_{p_i} A^{\rho_i^i-\eta_i-1} D \end{bmatrix}.$$

The vector $V \in \mathbb{R}^{\sum_{i=1}^{k^*} (p_i - \eta_i)}$ has the form:

$$V = \begin{bmatrix} \nu^1 \\ \vdots \\ \nu^{k^*} \end{bmatrix}, \nu^i = \begin{bmatrix} \nu \left(\bar{\chi}_{1, \rho_1^i}^i \right) \\ \vdots \\ \nu \left(\bar{\chi}_{p_i - \eta_i, \rho_{p_i - \eta_i}^i}^i \right) \end{bmatrix}$$

and is an available information that is computable online. Following the same arguments as in Proposition 1, one has $\text{rank}(\Theta^D) = m$. Thus, the relations (1.22) provide a finite time estimation \hat{w} of the unknown inputs w :

$$\hat{w} = (\Theta^D)^+ V$$

where $(\Theta^D)^+$ is the pseudo-inverse of Θ^D .

1.5 Example

By way of illustration, let us consider the system described by the following differential equations:

$$\dot{x} = \begin{bmatrix} -2 & -1 & 0 & 1 & 0 \\ 0 & -1 & 0 & 0 & 1 \\ 1 & 0 & -1 & 0 & 0 \\ 0 & -1 & -2 & 0 & 1 \\ 2 & 0 & 0 & 0 & -1 \end{bmatrix} x + \begin{bmatrix} 0 & 0 \\ 0 & -1 \\ 1 & 0 \\ 0 & -2 \\ -2 & 1 \end{bmatrix} w$$

$$y = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \end{bmatrix} x = \begin{bmatrix} C_1 \\ C_2 \end{bmatrix} x$$

Note that $\text{rank}(CD) \neq \text{rank}(D)$, and that the system can not be put in a form similar to (1.5) suitable for the design of classical step-by-step sliding mode observers. Applying Step 1 of the algorithm, one can check that $I_1 = 0$ and:

$$D_1 = \begin{bmatrix} C_1 \\ C_1 A \\ C_2 \end{bmatrix}$$

$$\Gamma_1 = \begin{bmatrix} C_1 A D \\ C_2 D \end{bmatrix} = \begin{bmatrix} 0 & -1 \\ 0 & -1 \end{bmatrix}.$$

Thus $A_1 = \begin{bmatrix} -1 & 1 \end{bmatrix}$ and one can choose the fictitious output as $y^2 = A_1 \begin{bmatrix} C_1 A^2 \\ C_2 A \end{bmatrix} x = C_1^2 x \in \mathbb{R}$. Step 2 implies that $I_2 = 0$ and $D_2 = \begin{bmatrix} C_1^2 \\ C_1^2 A \end{bmatrix}$. Since

$$\text{rank}(T) = \text{rank} \begin{bmatrix} D_1 \\ D_2 \end{bmatrix} = 5,$$

one can define the state transformation $z = Tx$. This gives

$$\dot{z} = \begin{cases} z_2 & \\ -4z_1 - 2z_2 + z_3 + 2z_4 + z_5 & -w_2 \\ -4z_1 - 2z_2 + z_3 + 3z_4 + z_5 & -w_2 \\ z_5 & \\ 6z_1 + 6z_2 - 2z_3 - 5z_4 - 4z_5 & -2w_1 + w_2 \end{cases}$$

$$y = [z_1 \ z_3]^T$$

Then, the observer is designed as follows:

$$\frac{d\hat{z}_1}{dt} = \nu(y_1 - \hat{z}_1) \quad (1.23)$$

$$\frac{d\hat{z}_2}{dt} = -4z_1 - 2\hat{z}_2 + z_3 + 2\hat{z}_4 + \hat{z}_5 + E(y_1 - \hat{z}_1) \nu(\tilde{z}_2 - \hat{z}_2) \quad (1.24)$$

$$\frac{d\hat{z}_3}{dt} = -4z_1 - 2\hat{z}_2 + z_3 + 3\hat{z}_4 + \hat{z}_5 + \nu(y_2 - \hat{z}_3) \quad (1.25)$$

$$\frac{d\hat{z}_4}{dt} = \nu(y^2 - \hat{z}_4) \quad (1.26)$$

$$\frac{d\hat{z}_5}{dt} = 6z_1 + 6\hat{z}_2 - 2z_3 - 5\hat{z}_4 - 4\hat{z}_5 + E(y^2 - \hat{z}_4) \nu(\tilde{z}_5 - \hat{z}_5) \quad (1.27)$$

The function E is defined by

$$E(s) = \begin{cases} 0 & \text{if } s \neq 0 \\ 1 & \text{otherwise} \end{cases}$$

and

$$\tilde{z}_2 = \nu(y_1 - \hat{z}_1)$$

$$\tilde{z}_5 = \nu(y^2 - \hat{z}_4)$$

Let us first consider the subsystems (1.23-1.24) and (1.25) (where the information injection is available), which are in block observable triangular form. The error dynamics $e_i = z_i - \hat{z}_i$ can be described by:

$$\dot{e}_1 = z_2 - \nu(e_1)$$

$$\dot{e}_2 = -2e_2 - 2e_4 + e_5 - w_2 - E(y_1 - \hat{z}_1) \nu(\tilde{z}_2 - \hat{z}_2)$$

$$\dot{e}_3 = -2e_2 - 3e_4 + e_5 - w_2 - \nu(e_3)$$

From the methodology given in section 1.4.1, one recover in a finite time t_1 the state variables z_1 , z_2 and z_3 and the equivalent output injections:

$$\begin{aligned}\nu(e_2) &= 2e_4 + e_5 - w_2 \\ \nu(e_3) &= 3e_4 + e_5 - w_2\end{aligned}$$

It can be seen that the quantity $y^2 - \hat{z}_4 = e_4 = \nu(e_3) - \nu(e_2)$ is henceforth available on the sliding manifold $\{e_2 = 0 \cap e_3 = 0\}$. Thus, after t_1 , the second part of the observer can be used to reconstruct the remaining variables:

$$\begin{aligned}\dot{e}_4 &= z_5 - \nu(e_4) \\ \dot{e}_5 &= 5e_4 - 4e_5 - 2w_1 + w_2 + E(y^2 - \hat{z}_4)\nu(\tilde{z}_5 - \hat{z}_5)\end{aligned}$$

Again, we have a step-by-step observer. Thus, the errors e_4 and e_5 converge to zero in finite time and one has the corresponding equivalent information injection

$$\nu(e_5) = 2w_1 - w_2.$$

The disturbance w is also estimated in finite time since:

$$\begin{aligned}w_1 &= -\frac{1}{2}(\nu(e_2) + \nu(e_5)) \\ w_2 &= -\nu(e_2) = -\nu(e_3)\end{aligned}$$

Simulations were performed for the observer (1.23-1.27). In practise, the function E is set as follows:

$$E(s) = \begin{cases} \varepsilon & \text{if } s \neq 0 \\ 1 & \text{otherwise} \end{cases}$$

where ε is a small positive constant. Figure 1.1 depicts the unmeasured states and their estimates. It can be seen that the observer converges quite fast and with good accuracy, as it can be seen in Figure 1(d) in the case of the error between x_5 and its estimate. Figure 1.2 illustrates the fact that the unknown inputs (that have been chosen as two sine functions) are recovered satisfactory.

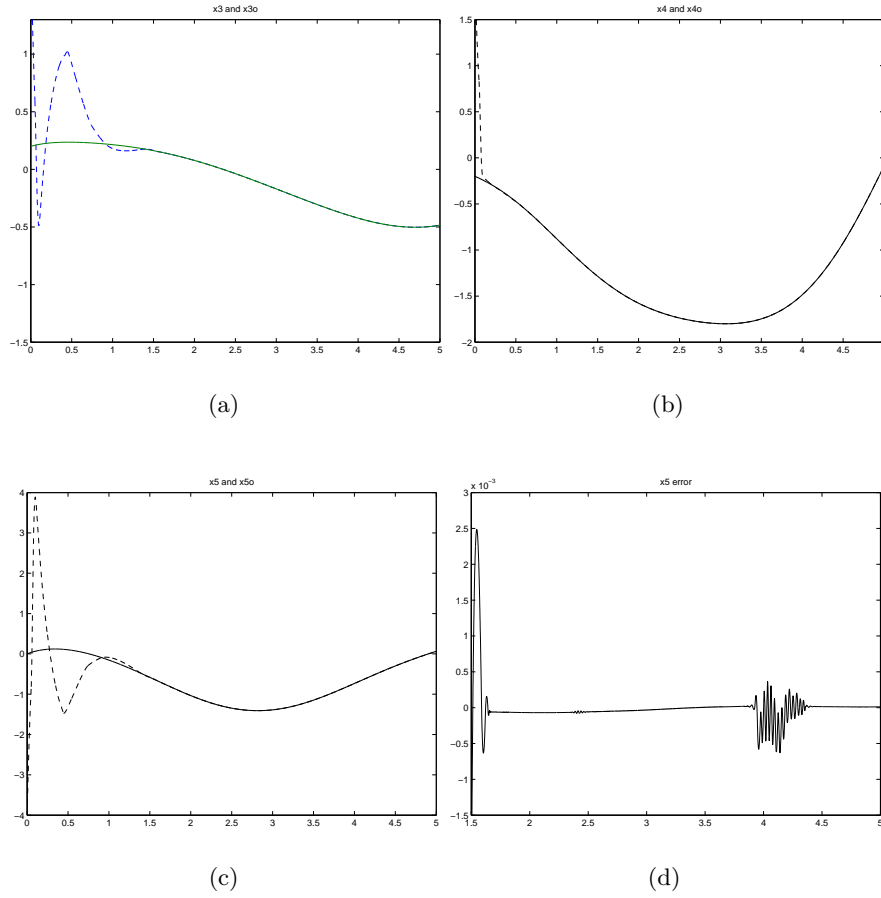


Fig. 1.1. State (solid line) and estimation (dotted line)

1.6 Concluding remarks

In this work, a new approach to solve the problem of designing an unknown input observer for linear systems has been developed. An algorithm was given in order to find a suitable change of coordinates for the design of a step-by-step second order sliding mode observer. This observer provides, by using the equivalent output injections, a finite time estimation of both state and unknown inputs. Since the observer is based on second order sliding mode algorithms, the equivalent output injections are obtained in a continuous way without any use of low pass filters.

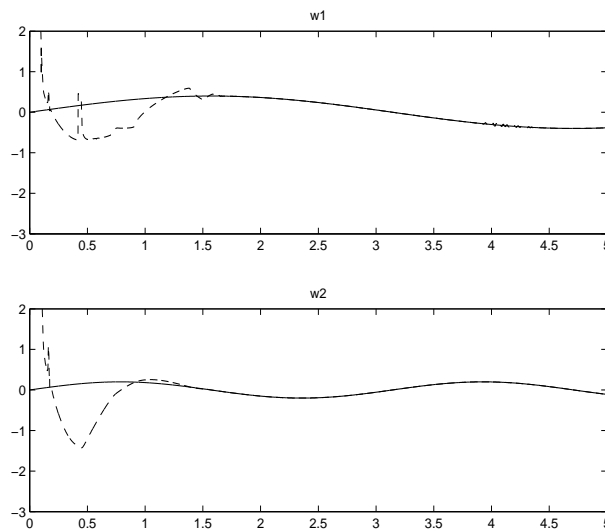


Fig. 1.2. Unknown input (solid line) and its estimation (dotted line)

It is important to note that this algorithm gives a solution to the observation problem of a class of systems subject to unknown inputs that do not necessarily satisfy the standard matching condition $\text{rank}(CD) = \text{rank}(D)$ as it has been seen in the example. It can also be said that the method given here extends other results in the area of sliding mode observers. Indeed, the first step of the algorithm takes into account existing methods that require that the systems can be put in a block observable form where the unknown inputs appear in the last block.

The Output Information Algorithm is a constructive one, and one has to go through all the algorithm to know if it can (or can not) provide a suitable change of coordinates for the design of a step-by-step sliding mode observer. The problem of finding structural conditions (if any) on the original system stating, a priori, whether this method might work or not is an open one. Future research is also concerned with the case of nonlinear systems and with practical applications, as for instance fault diagnosis.

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